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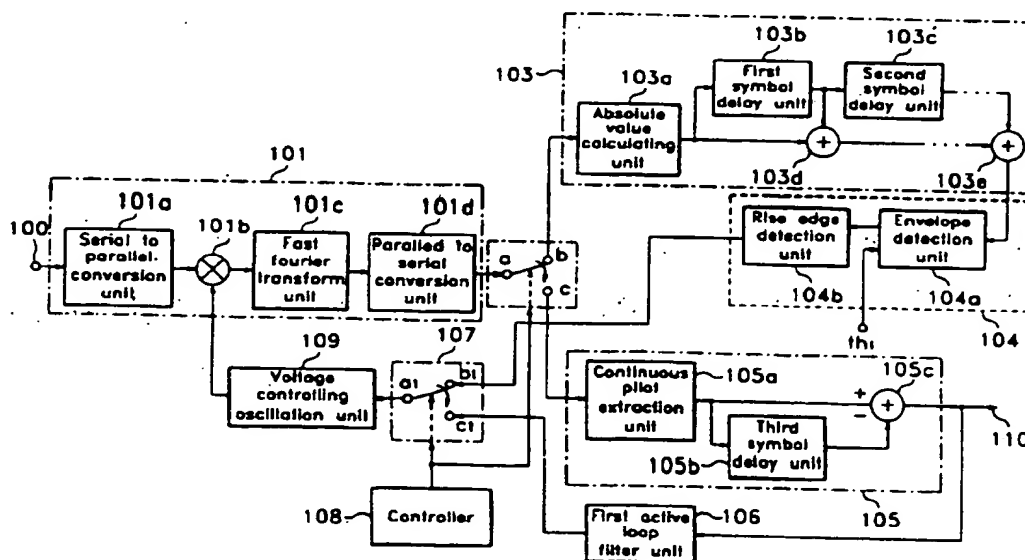
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(54) Abstract Title

## Synchronizing OFDM receivers

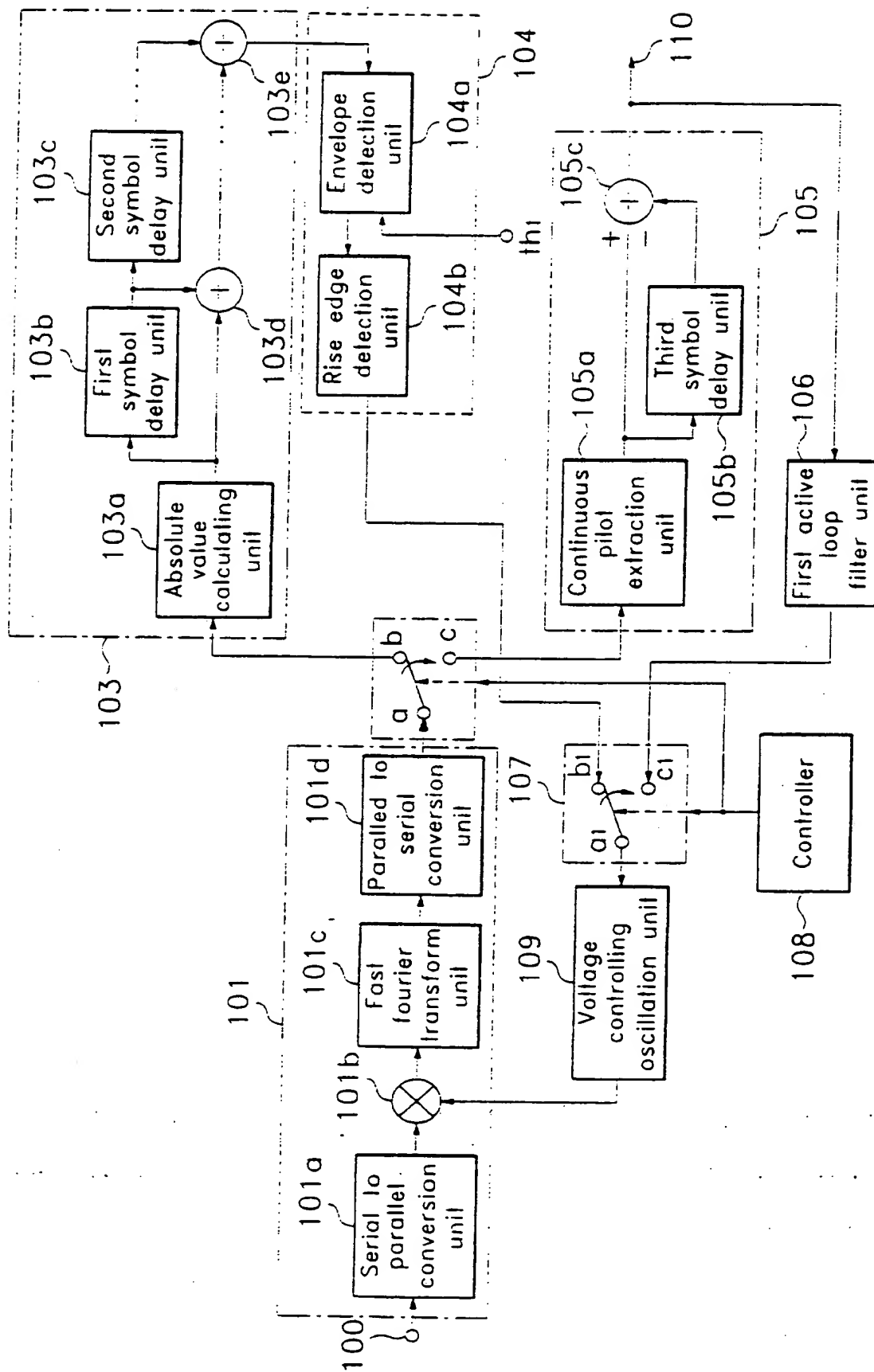
(57) The receiver of an OFDM transmission system is synchronized to compensate for carrier frequency offset by switching between: a first mode in which a local oscillator (109) in the receiver is coarsely adjusted by; calculating (103a) the signal strength of each received symbol, accumulating (103d... 103e) the signal strength of the received and appropriately delayed (103b, 103c...) signal strengths of earlier symbols, determining the timing (104b) at which the accumulated signal strength reaches a predetermined threshold value (th1) and coarsely adjusting the phase of the oscillator in accordance with that timing and; a second mode in which the local oscillator is finely corrected by; extracting (105a) a pilot signal from the received signals, calculating (105c) the phase difference between the extracted pilot (or carrier) and a previously extracted pilot (or carrier) signal delayed (105b) by one symbol duration and finely adjusting the phase of the oscillator in accordance with that phase difference.

FIG. 1



GB 2324 447 A

FIG. 1





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Claims searched: 1 to 24

Examiner: Ken Long  
Date of search: 12 August 1998

**Patents Act 1977**  
**Search Report under Section 17**

**Databases searched:**

UK Patent Office collections, including GB, EP, WO & US patent specifications, in:

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Other: ONLINE : WPI

**Documents considered to be relevant:**

Category	Identity of document and relevant passage	Relevant to claims
X	GB 2313527 A MITSUBISHI (page 3 lines 10-13, page 3 line 17 to page 4 line 11, page 6 lines 22-24 and page 7 lines 3-5)	1 & 10
X	GB 2313270 A MITSUBISHI (page 1 lines 6-9 and page 29 lines 13-22)	1 & 10
X	GB 2313022 A MITSUBISHI (page 1 lines 6-10 and 19, page 3 lines 4-16 and page 14 lines 19-22)	1 & 10

X	Document indicating lack of novelty or inventive step	A	Document indicating technological background and/or state of the art.
Y	Document indicating lack of inventive step if combined with one or more other documents of same category.	P	Document published on or after the declared priority date but before the filing date of this invention.
&	Member of the same patent family	E	Patent document published on or after, but with priority date earlier than, the filing date of this application.

directly by the present output of said offset detecting means and an integer gain for controlling input signal of the voltage-controlled oscillating means by averaging the previous outputs of said offset detecting means.

5

21. The apparatus of any of claims 11 to 20, further comprising a controller for controlling said first and second switch means according to the coarse mode and the fine mode.

10

22. The apparatus of Claim 21, wherein said controller decides said fine mode according to the accumulated value of the symbol of said symbol accumulating means and controls said first and second mode switch means.

15

23. A method for synchronizing a carrier frequency of an OFDM transmission system, the method being substantially as herein described with reference to Figure 1.

20

24. Apparatus for synchronizing the carrier frequency of an OFDM transmission system, the apparatus being substantially as herein described with reference to Figure 1.

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absolute value calculating means for calculating the signal strength of said data of carrier which is obtained by said first mode switch means;

5        symbol delay means for delaying the symbol successively as one symbol duration by considering said signal strength calculated by said absolute value calculating means; and

10        adding means for accumulating said strength value of the delayed signal by said symbol delay means and said strength value of the obtained signal by said absolute value calculating means and then outputting the accumulated value.

15

14. The apparatus of Claim 11, 12 or 13, wherein said hold value detection means comprises:

20        envelope detecting means for detecting an envelope by comparing said accumulated value of the signal strength obtained by said adding means with a predetermined threshold value; and

25        rise edge detection means for controlling said voltage-controlled oscillating means through said second mode switch means by calculating the detected envelope value by said envelope detecting means.

30        15. The apparatus of Claim 11, 12, 13 or 14, wherein said offset detection means comprises:

35        continuous pilot extracting means for extracting a continuous pilot signal from the serial data of carrier which is obtained by said first mode switch means;



second mode switch means for selecting and outputting one value out of the phase difference value which is filtered and then obtained by the loop filter means and the hold value obtained by the hold value detecting means;  
5 and

voltage-controlled oscillating means which is locked to the hold value or phase difference value which is selected by the second mode switch means for correcting  
10 the integer part and the prime part of the carrier frequency offset of the data demodulation means.

11. The apparatus of claim 10, further comprising control means for switching and controlling the first and second  
15 switch means with a predetermined time difference.

12. The apparatus of Claim 11, wherein said data demodulation means comprises:

20 serial to parallel conversion means for converting said inputted data of carrier into parallel data;

mixing means for mixing said converted parallel data and said oscillated voltage which is feedback by said  
25 voltage-controlled oscillating means;

fast Fourier transform means for extracting sample data by converting said mixed data of carrier in orthogonality by said mixing means; and  
30

parallel to serial conversion means for converting said sample data of carrier into serial data and outputting said data.

35 13. The apparatus of Claim 11 or 12, wherein said symbol accumulating means comprises:

predetermined Hz by controlling the gain proportionally with the calculated value of the phase difference.

10. An apparatus for a synchronizing carrier frequency of  
5 the OFDM transmission system, comprising:

data demodulation means for sampling a received  
signal by an oscillating feedback voltage from the  
obtained data of carrier by input terminal;  
10

first mode switch means for switching the demodulated  
carrier by said data demodulator means into a coarse mode  
or a fine mode for correcting an integer part or a prime  
part of a carrier frequency offset;  
15

symbol accumulating means for accumulating the  
calculated strength value of the frequency offset from the  
obtained data of carrier which is inputted by said first  
mode switch means and the strength value of the  
20 accumulated carrier frequency offset which is delayed by  
one symbol duration;

hold value detecting means for detecting a hold value  
by calculating the strength value of the accumulated  
25 carrier frequency offset until the value arrives at a  
threshold value;

offset detection means detecting the frequency offset  
from the demodulated carrier obtained by the first mode  
30 switch means and then obtaining the phase difference  
between the carrier frequency offset of the present symbol  
and that of the previous symbol;

loop filter means for controlling a gain of the  
35 obtained phase difference by the offset detecting means;

2. The method of Claim 1, wherein the signal strength of said carrier frequency is obtained by the square of an absolute value of the signal amplitude.

5 3. The method of Claim 1 or 2, wherein the correction of the integer part of said carrier frequency offset is performed within a predetermined frequency bandwidth between each carrier.

10 4. The method of Claim 2, wherein, after averaging said calculated signal strength of the carrier, the averaged value is set as an established value which is the comparative value of the signal strength of the carrier.

15 5. The method of Claim 3, wherein the correction of the integer part of said carrier frequency offset is performed by accumulating the calculated signal strength of the present carrier and the signal strength of a previous carrier during at least two symbol periods.

20 6. The method of Claim 3, wherein the integer part of said carrier frequency offset is corrected within a half frequency bandwidth of one subchannel.

25 7. The method of any of the preceding Claims, wherein said phase difference is obtained by dividing the value which is obtained by averaging the phase differences of pilots within one symbol by a predetermined value.

30 8. The method of Claim 7, wherein the phase difference between the adjacent symbols is obtained by dividing said average value of the phase differences of pilots within one symbol by  $2\pi_-(1+\Delta)$ .

35 9. The method of Claim 7, wherein the prime part of said carrier frequency offset is corrected within a

CLAIMS

1. A method for synchronizing a carrier frequency of the OFDM transmission system comprising, the steps of:

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determining which mode out of a coarse mode for correcting an integer part of a carrier frequency offset and a fine mode for correcting a prime part is selected;

10

calculating the strength of carrier which is received in each subchannel in the case of selecting the coarse mode;

15

accumulating the calculated strength of the carrier and the strength of a previous carrier which is delayed by one symbol duration for a predetermined period;

20

calculating the accumulated value of the strength of the carrier until the value arrives at a predetermined threshold value and correcting the integer part of the received carrier frequency offset using the calculated value;

25

extracting a pilot signal by receiving the carrier which is transmitted from a transmitter in the case of selecting the fine mode;

30

calculating a phase difference between the extracted pilot signal and a previous extracted pilot signal which is delayed by one symbol duration; and

35

correcting the prime part of the received carrier frequency offset within a predetermined frequency by controlling a gain of the calculated phase difference.

The reader's attention is directed to all papers and documents which are filed concurrently with or previous to this specification in connection with this application and which are open to public inspection with this  
5 specification, and the contents of all such papers and documents are incorporated herein by reference.

All of the features disclosed in this specification (including any accompanying claims, abstract and  
10 drawings), and/or all of the steps of any method or process so disclosed, may be combined in any combination, except combinations where at least some of such features and/or steps are mutually exclusive.

Each feature disclosed in this specification (including any accompanying claims, abstract and drawings), may be replaced by alternative features serving the same, equivalent or similar purpose, unless expressly  
15 stated otherwise. Thus, unless expressly stated otherwise, each feature disclosed is one example only of a generic series of equivalent or similar features.  
20

The invention is not restricted to the details of the foregoing embodiment(s). The invention extends to any  
25 novel one, or any novel combination, of the features disclosed in this specification (including any accompanying claims, abstract and drawings), or to any novel one, or any novel combination, of the steps of any method or process so disclosed.

30

oscillating unit 109 through the second mode switch unit 107. At this time, the gain of the voltage-controlled oscillating unit 109 is one (1).

5       The voltage-controlled oscillating unit 109 can reduce the prime part of the carrier frequency offset which is outputted from the mixing unit 101b within a several Hz by the gain of the first-order active loop filter unit 106.

10

Here, the natural frequency  $\omega_n$  should be decided as that the remaining carrier frequency offset is small and that it is possible to track the carrier in fast speed.

15       As described above, by moving both the coarse mode for compensating the integer part of the carrier frequency offset and the fine mode for compensating the prime part with a predetermined time difference, the synchronization of the carrier frequency in the OFDM transmission  
20       apparatus can be achieved.

While there have been illustrated and described what are considered to be preferred embodiments of the present invention, it will be understood by those skilled in the  
25       art that various changes and modifications may be made, and equivalents may be substituted for elements thereof without departing from the true scope of the present invention. In addition, many modifications may be made to adapt a particular situation to the teaching of the  
30       present invention without departing from the central scope thereof. Therefore, it is intended that the present invention not be limited to the particular embodiment disclosed as the best mode contemplated for carrying out the present invention, but that the present invention  
35       includes all embodiments falling within the scope of the appended claims.

The subtracting unit 105c calculates the phase difference between the present pilot signal extracted by the continuous pilot extracting unit 105a and the pilot signal which is transmitted and extracted in the same subchannel as the previous pilot signal which is delayed by the third symbol delay unit 105b.

In other words, by averaging the phase differences of the pilots within one symbol through the continuous pilot extracting unit 105a, the third symbol delay unit 105b and the subtracting unit 105c, a presumed value regarding the carrier frequency offset can be set. The phase difference of the pilots which is transmitted in the same subchannel between the adjacent two symbols is proportional to the carrier frequency offset. By dividing this averaged value by  $2\pi_-(1+\Delta)$ , the phase difference between the symbols can be obtained.

The phase difference obtained by the subtracter 105c is outputted through an output terminal 110 and filtered through the first-order active loop filter unit 106, thereby supplied to the second mode switch unit 107.

The first-order active loop filter unit 106 uses both a proportional gain  $K_p$  for input signal of the voltage-controlled oscillating unit 109 directly by the present output of the offset detecting unit 105 and an integer gain  $K_i$  for input signal of the voltage-controlled oscillating unit 109 by averaging the previous outputs of the offset detecting unit 105.

Here, the gains can be expressed as a natural frequency  $\omega_n$ , the proportional gain  $K_p$  becomes  $\omega_n(1+\omega_n)$ , and the integer gain  $K_i$  becomes  $4\omega_n^2$ .

35

The signal which passes the first-order active loop filter unit 106 is inputted to the voltage-controlled

subchannel and N is the number of points which are used in the fast Fourier transform unit 101c.

5 In units for the fine mode, there are the offset detecting unit 105, the first-order active loop filter unit 106 and the voltage-controlled oscillating unit 109.

10 The offset detecting unit 105 performs the fine mode using the fast Fourier transform unit 101c, the parallel to serial conversion unit 101d, the continuous pilot extracting unit 105a, the third symbol delay unit 105b and the subtracting unit 105c.

15 The operation of the fine mode for correcting the prime part is illustrated. First, when the operation of the coarse mode is completed, the controller 108 converts the movable terminals a and a1 of the first and second mode switch units 102 and 107 such as the analog switch to fixed branches c and c1.

20 Accordingly, the fast Fourier transform unit 101b performs the fast Fourier transform of the sampled sequence by N number of points, and then supplies it to the continuous pilot extracting unit 105a of the offset  
25 detecting unit 105 through the parallel to serial conversion unit 101d, the movable terminal a and fixed branch c of the first switch unit 102.

30 The continuous pilot extracting unit 105a of the offset detecting unit 105 extracts the pilot components continuously among the data of carrier of N number of points which are inputted through the fast Fourier transform unit 101c, and then supplies it to the subtracting unit 105c through the third symbol delay unit  
35 105b.



detecting unit 104a becomes larger than the threshold value th1. When the counting is completed at the rise edge point, the counted value is maintained and supplied to a fixed branch b1 of the second mode switch unit 107 such as  
5 the analog switch.

The second mode switch unit 107 is controlled together with the first mode switch unit 102 and is connected to the fixed branch b1 by the controller 108.  
10

As a result, the hold value which is detected by the rise edge detecting unit 104b is supplied to the voltage-controlled oscillating unit 109 through the second mode switch unit 107.  
15

The voltage-controlled oscillating unit 109 generates the oscillating voltage corresponding to the hold value which is inputted by the second mode switch unit 107 and is supplied it to the mixing unit 101b of the data  
20 demodulation unit 101. Here, the integer part of the carrier frequency offset is corrected within a half frequency bandwidth of one channel.

A timing control signal which is used in the fine mode is decided according to how many symbol durations the data of carrier is accumulated by the symbol accumulating unit 103.  
25

Moreover, in the fine mode, the carrier frequency offset is corrected within a several Hz by using the phase difference between the continuous pilot of the symbols with the carrier frequency offset.  
30

In other words, the carrier frequency offset( $\omega$ ) brings about the phase shift as much as  $2\pi \omega (1+\Delta/N)$  to the receiving signal, where  $\Delta$  is the frequency bandwidth of a  
35

The signal strength which is obtained by the absolute value calculating unit 103a is delayed in the first and second symbol delay units 103b and 103c each by one symbol unit. The signal strength which is delayed by one symbol  
5 by the first symbol delay unit 103b and the strength value of the present signal which is inputted by the absolute value calculating unit 103a are accumulated through the first adding unit 103d. The strength value of the  
10 accumulated signal is accumulated to the signal strength which is delayed by the second symbol delay unit 103c through the second adding unit 103e, and then it is supplied to the hold value detecting unit 104.

As there is a guard band whose signal power is zero  
15 (0) in the data of carrier which is switched by the first mode switch unit 102 to take a guard margin against the mutual interference by adjacent subchannels, the data of carrier regarding carrier indexes for several symbols are accumulated and then they are supplied to the hold value  
20 detecting unit 104.

The hold value detecting unit 104 calculates the strength of the carrier accumulated by the symbol accumulating unit 103 until it reaches the threshold value  
25  $th_1$ , and checks how much the signal of the guard band is shifted. After that, the hold value detecting unit 104 corrects the integer part of the carrier frequency offset within a half band of one channel. In other words, the  
30 envelope detecting unit 104a of the hold value detecting unit 104 detects the envelope and compare the signal strength of the carrier accumulated by the symbol accumulating unit 103 with the threshold value  $th_1$ . Here, the envelope is obtained by designating the threshold value  $th_1$  as the half of the signal power.

35 Moreover, the rise edge detecting unit 104b performs counting until the value obtained by the envelope

And then the fast Fourier transform unit 101c of the data demodulation unit 101 converts the inputted data of carrier through the mixing unit 101b orthogonally at a high speed and then outputs the Fourier-transformed data.

5

At this time, in the case that there is the carrier frequency offset in the data of carrier which is inputted serially in the mixing unit 101b, when the data demodulation unit 101 performs the orthogonal transformation of the data at a high speed, the phase of data is shifted circularly as much as the integer part value of the carrier frequency offset.

10

The data of carrier which is transformed orthogonally through the fast Fourier transform unit 101c is converted serially through the parallel to serial conversion unit 101d and it is supplied to a movable terminal a of the first mode switch unit 102 such as an analog switch.

15

At this time, when the controller 108 attaches the movable terminal a to a fixed branch b by controlling the first mode switch unit 102 to perform the coarse mode, the data of carrier which is converted by the parallel to serial conversion unit 101d is supplied to the symbol accumulating unit 103 through the movable branch a and the fixed branch b of the first mode switch unit 102.

20

25

The symbol accumulating unit 103 calculates the strength of the integer part of the carrier frequency offset from the data of carrier which is currently inputted and accumulates the calculated value and the strength of the carrier frequency offset which is delayed by one symbol during several symbol periods, and thereby the accumulated value is supplied to the hold value detecting unit 104.

30

35

oscillating unit 109 through the second mode switch unit 107 by calculating the detected envelope value by the envelope detecting unit 104a.

5           Moreover, the offset detecting unit 105 includes: a  
continuous pilot extracting unit 105a for extracting a  
continuous pilot signal from the serial data of carrier  
which is obtained by the first mode switch unit 102; a  
third symbol delay unit 105b for delaying the extracted  
10 continuous pilot signal by one symbol period; and a  
subtracting unit 105c for obtaining the phase difference  
between the continuous pilot signal which is delayed  
during one symbol period and a present continuous pilot  
signal and then supplying to the first-order active loop  
15 filter unit 106.

Referring to Figure 1, the operation of the  
embodiment of the present invention will be illustrated in  
detail.

20

First, when the data of carrier is inputted through  
the input terminal 100, the data demodulation unit 101  
samples the received signal by an oscillating feedback  
voltage which is obtained by the voltage-controlled  
25 oscillation unit 109, and then supplies the sampled data  
to the first mode switch unit 102.

In other words, the data of carrier which is inputted  
serially through the input terminal 100 is converted into  
30 the parallel data by the serial to parallel conversion  
unit 101a of the data demodulation unit 101, and it is  
mixed with the oscillating voltage which is inputted by  
the voltage-controlled oscillation unit 109 in the mixing  
unit 101b, and then it is supplied to the fast Fourier  
35 transform unit 101c.

switching and controlling the first and second switch units 102 and 107 with a predetermined time difference.

5       The data demodulation unit 101 includes: a serial to  
parallel conversion unit 101a for converting the inputted  
data of carrier by the input terminal 100 into parallel  
data; a mixing unit 101b for mixing the converted parallel  
data by the serial to parallel conversion unit 101a and  
10   the oscillated voltage which is feedback by the voltage-  
controlled oscillating unit 109; a fast Fourier transform  
unit 101c for extracting a sample data by converting the  
mixed data of carrier by the mixing unit 101b in  
orthogonality; and a parallel to serial conversion unit  
101d for converting the sample data of carrier which is  
15   inputted by the fast Fourier transform unit 101c into  
serial data and outputting the data.

      Moreover, the symbol accumulating unit 103 includes:  
an absolute value calculating unit 103a for calculating  
20   the signal strength of the data of carrier which is  
obtained by the first mode switch unit 102; first and  
second symbol delay units 103b and 103c for successively  
delaying the symbol as one symbol unit by considering the  
signal strength calculated by the absolute value  
25   calculating unit 103a; and first and second adding units  
103d and 103e for accumulating the strength value of the  
delayed signal by the first and second symbol delay units  
103b and 103c and the strength value of the signal which  
is obtained by the absolute value calculating unit 103a  
30   and then outputting the accumulated value.

      The hold value detecting unit 104 includes: an  
envelope detecting unit 104a for detecting an envelope by  
comparing the accumulated value of the signal strength  
35   obtained by the second adding unit 103e with a  
predetermined threshold value th1; and a rise edge  
detecting unit 104b for controlling the voltage-controlled

(OFDM) transmission system according to the present invention.

Referring to Figure 1, the apparatus for a  
5 synchronizing carrier frequency of the OFDM transmission  
system includes: a data demodulation unit 101 for sampling  
a received signal by an oscillating feedback voltage from  
the obtained data of carrier by input terminal 100; a  
10 first mode switch unit 102 for switching the demodulated  
carrier by the data demodulator 101 into a coarse mode or  
a fine mode; a symbol accumulating unit 103 for  
accumulating the calculated strength value of the  
frequency offset from the obtained data of carrier which  
15 is inputted by the first mode switch unit 102 and the  
strength value of the accumulated carrier frequency offset  
which is delayed by one symbol duration; a hold value  
detecting unit 104 for detecting a hold value by  
calculating the strength value of the accumulated carrier  
20 frequency offset until the value arrives at a threshold  
value; an offset detecting unit 105 for detecting the  
frequency offset from the demodulated carrier obtained by  
the first mode switch unit 102 and then obtaining the  
phase difference between the carrier frequency offset of  
25 this symbol and that of the previous symbol; a first-order  
active loop filter unit 106 for controlling the gain of  
the obtained phase difference by the offset detecting unit  
105; a second mode switch unit 107 for selecting and  
outputting one value out of the phase difference value  
30 which is filtered and then obtained by the first-order  
active loop filter unit 106 and the hold value obtained by  
the hold value detecting unit 104; and a voltage-  
controlled oscillating unit 109 which is locked to the  
hold value or phase difference value which is selected by  
35 the second mode switch unit 107 for correcting the integer  
part and the prime part of the carrier frequency offset of  
the data demodulation unit 101; and a control unit 108 for

signal by one symbol period; and a subtracting unit for obtaining the phase difference between the continuous pilot signal which is delayed during one symbol period and a present continuous pilot signal.

5

As a result, it is possible to exactly correct the integer part of carrier frequency offset in the coarse mode for compensating the integer part of the carrier frequency offset and in the fine mode for compensating the prime part, with a predetermined time difference. Accordingly, even in the case that the carrier frequency offset is above the bandwidth of one subchannel, the integer part and the prime part of the carrier frequency offset can be corrected, thereby resulting in the exact synchronization of carrier frequency.

15

A more complete appreciation of the present invention, and many of the attendant advantages thereof, will become readily apparent as the same becomes better understood by reference to the following detailed description when considered in conjunction with the accompanying drawings in which like reference symbols indicate the same or similar components, wherein:

20

Figure 1 is a block diagram illustrating an embodiment of a synchronizing apparatus of carrier frequency in an OFDM transmission system according to the present invention.

25

The objects, characteristics and advantages of the above-described invention will be more clearly understood through the preferable embodiments referring to the attached drawings.

30

Figure 1 is a block diagram illustrating a preferred embodiment of apparatus for a synchronizing carrier frequency of an orthogonal frequency division multiplexing

35

parallel data; a mixing unit for mixing the converted  
parallel data by the serial to parallel conversion unit  
and the oscillated voltage which is feedback by the  
voltage-controlled oscillating unit; a fast Fourier  
5 transform unit for extracting a sample data by converting  
the mixed data of carrier in orthogonality; and a parallel  
to serial conversion unit for converting the sample data  
of carrier into serial data and outputting the data.

10 Preferably, the symbol accumulating unit includes: an  
absolute value calculating unit for calculating the signal  
strength of the data of carrier which is obtained by the  
first mode switch unit; at least one symbol delay unit for  
successively delaying the symbol as one symbol unit by  
15 considering the signal strength calculated by the absolute  
value calculating unit; and an adding unit for  
accumulating the strength value of the delayed signal by  
the symbol delay unit and the strength value of the signal  
obtained by the absolute value calculating unit and then  
20 outputting the accumulated value.

The hold value detecting unit preferably includes: an  
envelope detecting unit for detecting an envelope by  
comparing the accumulated value of the signal strength  
25 obtained by the adding unit with a predetermined threshold  
value; and a rise edge detecting unit for controlling the  
voltage-controlled oscillating unit through the second  
mode switch unit by calculating the detected envelope  
value.

30 Preferably, the offset detecting unit includes: a  
continuous pilot extracting unit for extracting a  
continuous pilot signal from the serial data of carrier  
which is obtained by the first mode switch unit; a symbol  
35 delay unit for delaying the extracted continuous pilot



Preferably, at least two said symbol delay means and adding means are provided and accumulates the signal strength of the carrier during at least two symbols.

5        Said threshold value may be obtained by averaging the signal strengths of the carrier and compared with the value which is accumulated during a predetermined symbol.

10        Said rise edge detection means preferably performs counting until the envelop value detected by said envelop detection means becomes larger than said threshold value and then supplies said counted value said voltage-controlled oscillating means.

15        Said loop filter means preferably comprise a first-order active loop filter having both a proportional gain for controlling the input signal of said voltage-controlled oscillating means directly by the present output of said offset detecting means and an integer gain  
20        for controlling input signal of the voltage-controlled oscillating means by averaging the previous outputs of said offset detecting means.

25        The apparatus preferably comprises a controller for controlling said first and second switch means according to the coarse mode and the fine mode.

30        Said controller may decide said fine mode according to the accumulated value of the symbol of said symbol accumulating means and controls said first and second mode switch means.

35        The data demodulation unit preferably includes: a serial to parallel conversion unit for converting the inputted data of carrier by the input terminal into

by said first mode switch means; symbol delay means for delaying the symbol successively as one symbol duration by considering said signal strength calculated by said absolute value calculating means; and adding means for  
5 accumulating said strength value of the delayed signal by said symbol delay means and said strength value of the obtained signal by said absolute value calculating means and then outputting the accumulated value.

10 Said hold value detection means preferably comprises: envelope detecting means for detecting an envelope by comparing said accumulated value of the signal strength obtained by said adding means with a predetermined  
15 threshold value; and rise edge detection means for controlling said voltage-controlled oscillating means through said second mode switch means by calculating the detected envelope value by said envelope detecting means.

Said offset detection means may comprise: continuous  
20 pilot extracting means for extracting a continuous pilot signal from the serial data of carrier which is obtained by said first mode switch means; symbol delay means for delaying and outputting said extracted continuous pilot signal by one symbol period; and subtracting means for  
25 obtaining the phase difference between said continuous pilot signal which is delayed during one symbol period and a present continuous pilot signal and then supplying to said loop filter means.

30 Said absolute value calculating means may obtain the signal strength of said carrier by squaring an absolute value.

value; an offset detecting unit for detecting the frequency offset from the demodulated carrier obtained by the first mode switch unit and then obtaining the phase difference between the carrier frequency offset and the carrier frequency offset of the previous symbol; a loop filter unit for controlling the gain of the obtained phase difference; a second mode switch unit for selecting and outputting one value out of the phase difference value which is filtered and then obtained and the hold value obtained by the hold value detecting unit; and a voltage-controlled oscillating unit which is locked to the hold value or phase difference value which is selected by the second mode switch unit for correcting the integer part and the prime part of the carrier frequency offset of the data demodulation unit.

The apparatus preferably includes control means for switching and controlling the first and second switch means with a predetermined time difference.

Said data demodulation means preferably comprises: serial to parallel conversion means for converting said inputted data of carrier into parallel data; mixing means for mixing said converted parallel data and said oscillated voltage which is feedback by said voltage-controlled oscillating means; fast Fourier transform means for extracting sample data by converting said mixed data of carrier in orthogonality by said mixing means; and parallel to serial conversion means for converting said sample data of carrier into serial data and outputting said data.

Said symbol accumulating means preferably comprises: absolute value calculating means for calculating the signal strength of said data of carrier which is obtained

phase differences of pilots within one symbol by a predetermined value.

5 The phase difference between the adjacent symbols is preferably obtained by dividing said average value of the phase differences of pilots within one symbol by  $2\pi_{-}$   $(1+\Delta)$ .

10 The prime part of said carrier frequency offset may be corrected within a predetermined Hz by controlling the gain proportionally with the calculated value of the phase difference.

15 Preferably, the threshold value is set to the half the strength value of the received carrier.

20 Preferably, the phase difference is obtained by dividing the averaged value of the phase differences between each pilot and its previous pilot within one symbol duration by a predetermined value.

According to another aspect of the present invention, an apparatus for synchronizing the carrier frequency of the OFDM transmission system includes: a data demodulation  
25 unit for sampling a received signal by an oscillating feedback voltage from the obtained data of carrier; a first mode switch unit for switching the demodulated carrier into a coarse mode and a fine mode; a symbol accumulating unit for accumulating the calculated strength  
30 value of the frequency offset from the obtained data of carrier and the strength value of the accumulated carrier frequency offset which is delayed by one symbol duration; a hold value detecting unit for detecting a hold value by calculating the strength value of the accumulated carrier  
35 frequency offset until the value arrives at a threshold

receiving the carrier which is transmitted from a transmitter in the case of selecting the fine mode; calculating a phase difference between the extracted pilot signal and a previous extracted pilot signal which is  
5 delayed by one symbol duration; and correcting the prime part of the received carrier frequency offset within a predetermined frequency by controlling a gain of the calculated phase difference.

10        Preferably, the signal strength of said carrier frequency is obtained by the square of an absolute value of the signal amplitude.

15        The correction of the integer part of said carrier frequency offset may be performed within a predetermined frequency bandwidth between each carrier.

20        After averaging said calculated signal strength of the carrier, the averaged value is preferably set as an established value which is the comparative value of the signal strength of the carrier.

25        The correction of the integer part of said carrier frequency offset is preferably performed by accumulating the calculated signal strength of the present carrier and the signal strength of a previous carrier during at least two symbol periods.

30        The integer part of said carrier frequency offset is preferably corrected within a half frequency bandwidth of one subchannel.

Said phase difference is preferably obtained by dividing the value which is obtained by averaging the

however, above the bandwidth of one subchannel, it is difficult to measure the rotational amount of the phase, and thereby the synchronization is also failed.

5           Therefore, it is an aim of preferred embodiments of the present invention to provide a method and apparatus for synchronizing a carrier frequency of orthogonal frequency division multiplexing (OFDM) transmission apparatus capable of successfully synchronizing the  
10 carrier frequency even in the case that the offset of the carrier frequency is above the bandwidth of one subchannel.

          It is another aim to synchronize the carrier  
15 frequency by cooperating both a coarse mode for compensating an integer part of a carrier frequency offset and a fine mode for compensating a prime part of the carrier frequency offset, in turn with a predetermined time interval.

20           According to one aspect of the present invention, a method for synchronizing a carrier frequency of the OFDM transmission system includes the steps of: determining which mode out of a coarse mode for correcting an integer  
25 part of a carrier frequency offset and a fine mode for correcting a prime part is selected; calculating the strength of carrier which is received in each subchannel in the case of selecting the coarse mode; accumulating the calculated strength of the carrier and the strength of a  
30 previous carrier which is delayed by one symbol duration for a predetermined period; calculating the accumulated value of the strength of the carrier until the value arrives at a predetermined threshold value and correcting the integer part of the received carrier frequency offset  
35 using the calculated value; extracting a pilot signal by

is divided into an integer part and a prime part through a normalization by the frequency bandwidth of one subchannel.

5       The integer part of the offset causes the signals by the fast Fourier transform (FFT) to be circularly shifted at a receiver, and its prime part causes the power and phase of the signals to be disturbed by an interference with the subchannels.

10

As described above, when the carrier frequency is offset between a transmitter and a receiver, as a result, the offset of the carrier frequency causes the interference between the carriers of subchannels, and the orthogonality between the subchannels cannot be maintained, thereby increasing the error rate of transmitted data.

15

In the OFDM transmission systems, for methods which can be used in a terrestrial high definition television, a 8K discrete Fourier transform (DFT) and a 2K discrete Fourier transform are standardized. The recovery of the carrier has been developed as the main technology of the transmission.

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25

The existing methods for synchronization used in the OFDM transmission system, synchronize the carrier frequency by using the fact that its phase is rotated uniformly in the case that the carrier frequency is not synchronized.

30

The conventional methods for synchronization in the OFDM transmission apparatus works well when the carrier frequency offset is within the bandwidth of one subchannel. When the carrier frequency offset is,

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METHOD AND APPARATUS FOR SYNCHRONIZING A CARRIER FREQUENCY  
OF AN ORTHOGONAL FREQUENCY DIVISION MULTIPLEXING  
TRANSMISSION SYSTEM

5           The present invention relates to a  
synchronization of carrier frequency in an orthogonal  
frequency division multiplexing (OFDM) transmission  
apparatus, and particularly to a method and apparatus for  
10       synchronizing a carrier frequency of an orthogonal  
frequency division multiplexing (OFDM) transmission  
apparatus capable of synchronizing a carrier frequency  
efficiently even if a carrier frequency offset is above  
the frequency bandwidth of one subchannel by cooperating  
both a coarse mode for synchronizing an integer part of  
15       the carrier frequency offset and a fine mode for  
synchronizing a prime part of the carrier frequency  
offset when performing the synchronization of carrier  
frequency in the OFDM transmission apparatus.

20           Generally, a digital television adopts a frequency  
division multiplexing (FDM) transmission system which is  
one of the multiple carrier modulation methods. It  
occupies an almost equal frequency bandwidth required in  
transmitting data a single carrier, since the data have  
25       the longer symbol period by being transmitted in parallel  
on several subchannels. The FDM signal is generated by an  
inverse fast Fourier transform (IFFT) and its modulation  
in parallel by carriers which are orthogonal to each  
other.

30           At this time, a synchronization of carrier frequency  
should be achieved between a transmitter and a receiver  
before performing a fast Fourier transform at the  
receiver. When the carrier frequency is offset between a  
35       transmitter and a receiver, the carrier frequency offset